

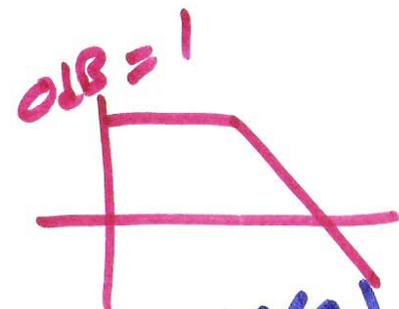
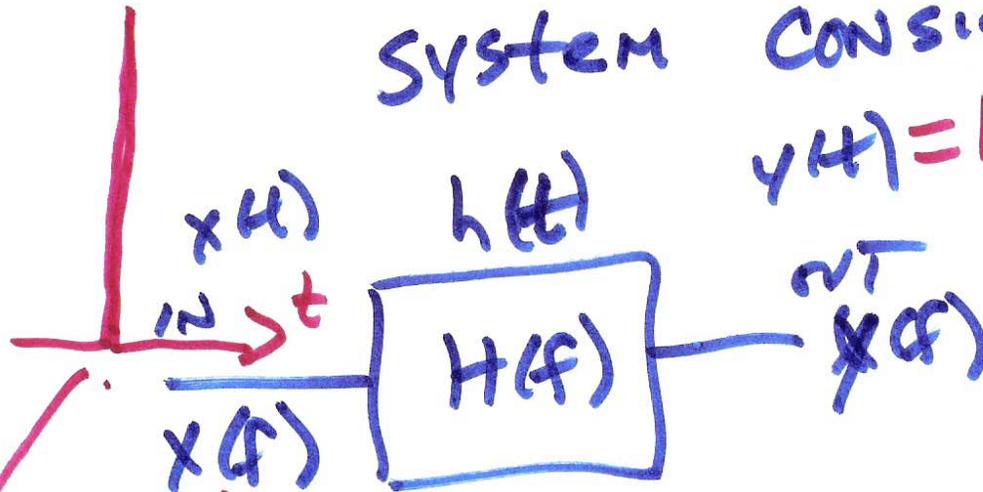
April 18, 2011

Lecture 21

SYSTEM Considerations

$y(t) = h(t)$ $h(t) =$ impulse response

$$\mathcal{F}\{h(t)\} = H(f)$$



$$H(f) = \frac{y(f)}{x(f)} = \frac{1}{1 + j2\pi fRC}$$

$$H(f) = \frac{y(f)}{x(f)}$$

1)

Distortionless TRANSMISSION.



$$e^{j\theta} = \cos \theta + j \sin \theta$$
$$|a + jb| = \sqrt{a^2 + b^2}$$
$$\angle a + jb = \tan^{-1} \frac{b}{a}$$

$$y(t) = x(t - t_0) \cdot K$$

$$Y(f) = X(f) \cdot e^{-j2\pi f \cdot t_0} \cdot K$$

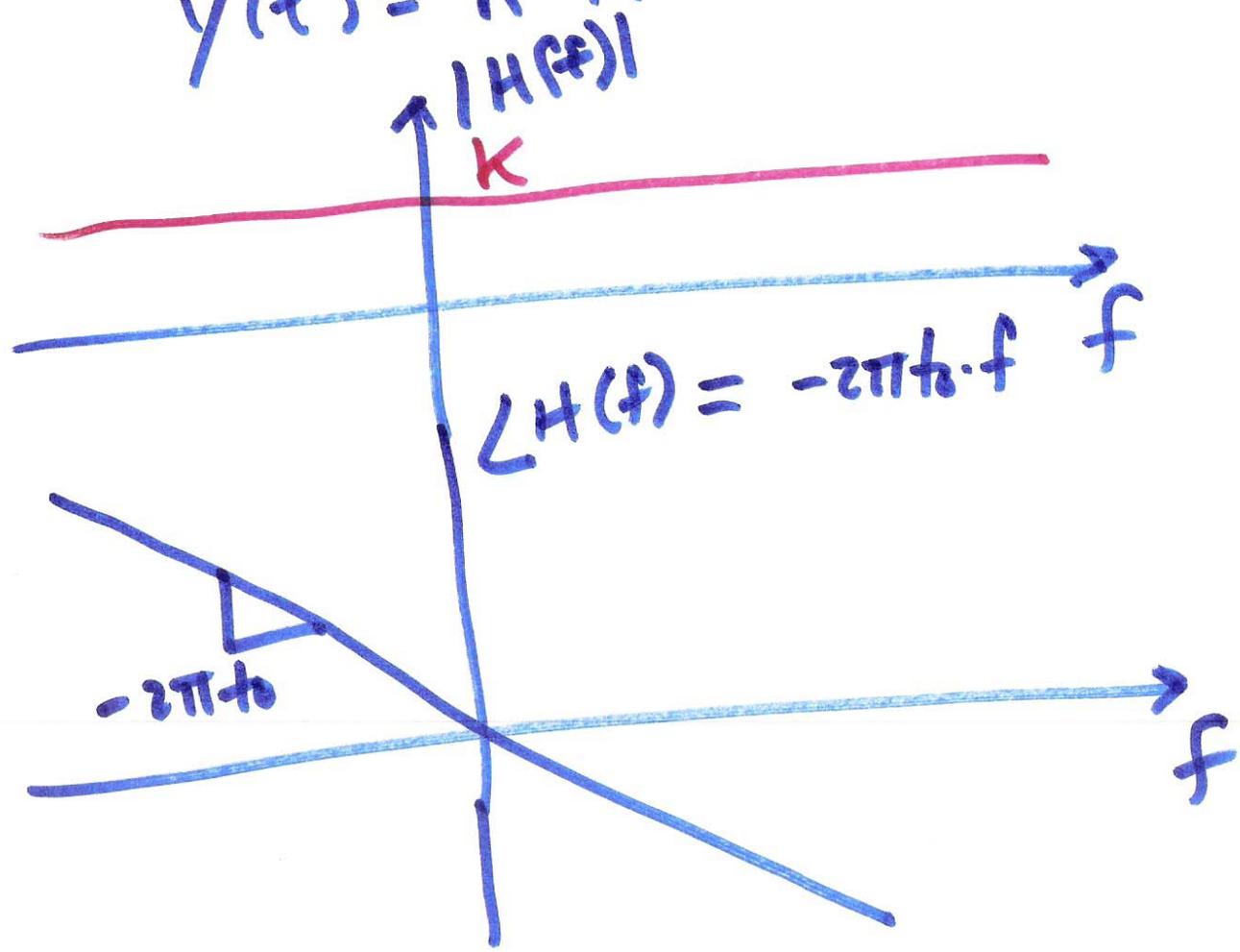
$$e^{-j2\pi f \cdot t_0} = \cos(-2\pi f t_0) + j \sin(-2\pi f t_0)$$

$$|e^{-j2\pi f t_0}| = \sqrt{\cos^2(-\pi f t_0) + \sin^2(-\pi f t_0)} = 1$$

2)

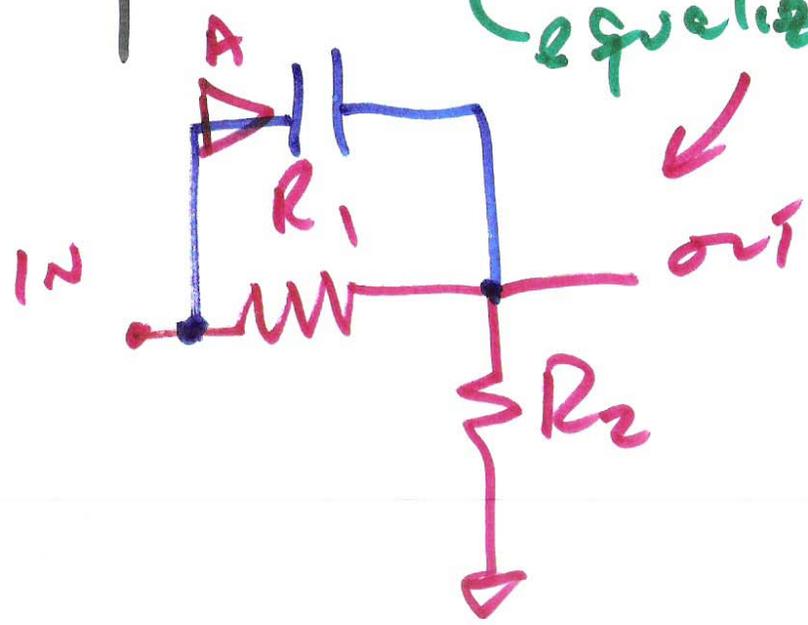
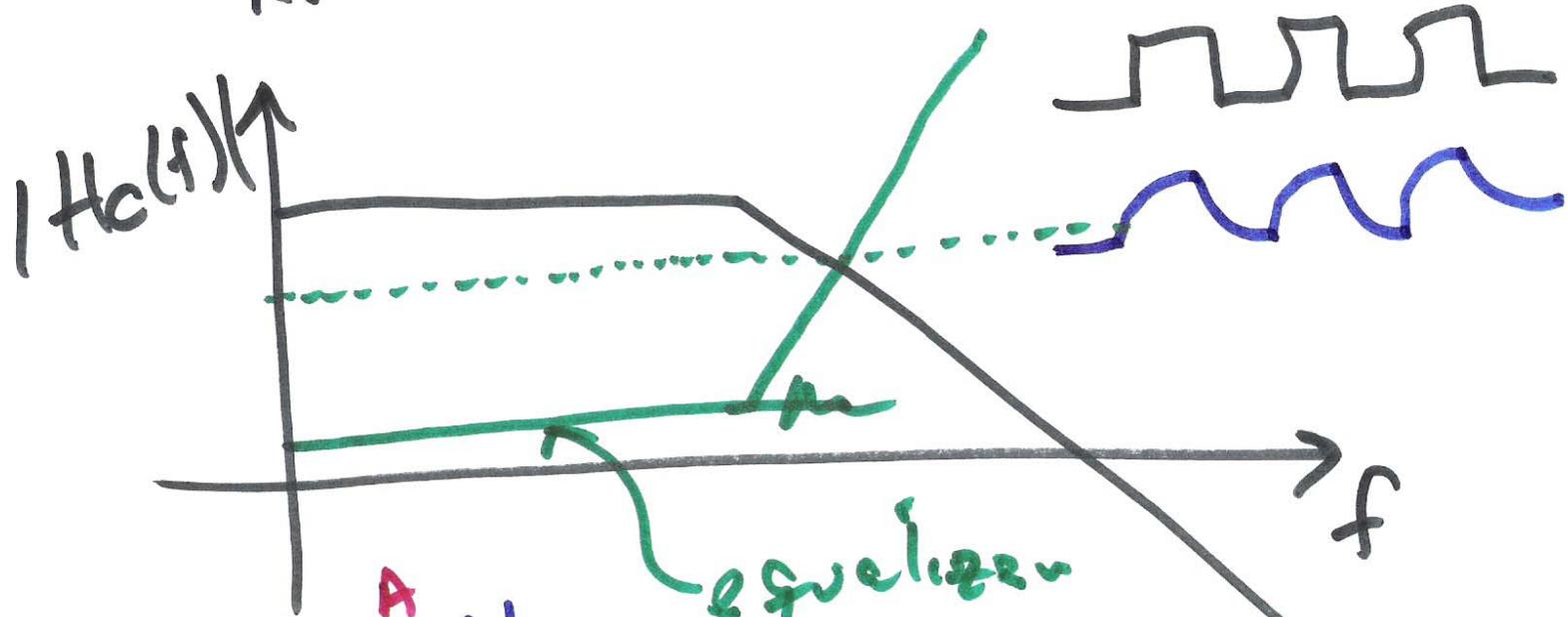
$$\tan^{-1} \frac{\sin(-2\pi f t_0)}{\cos(-2\pi f t_0)} = -2\pi f t_0 = \angle \tan(-2\pi f t_0)$$

$$y(t) = k \cdot x(t - t_0)$$



3)

Real Communications Channel



4)

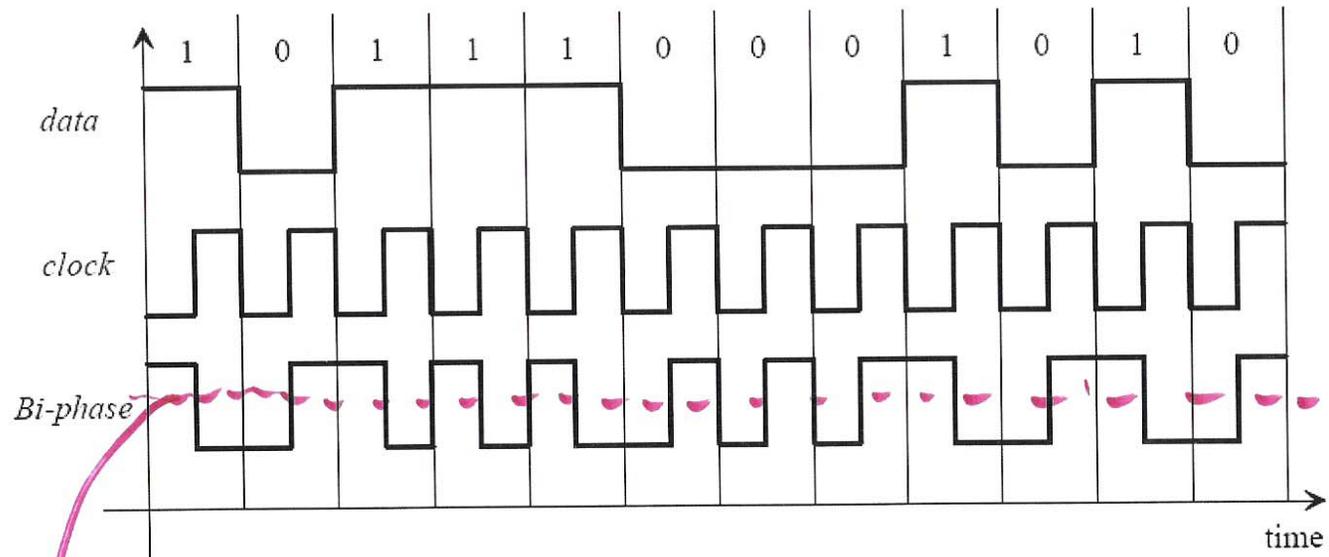
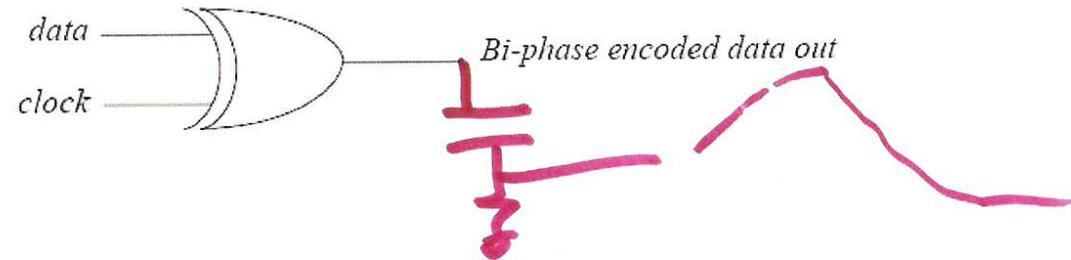


Figure 19.43 Bi-phase data encoding.



5)

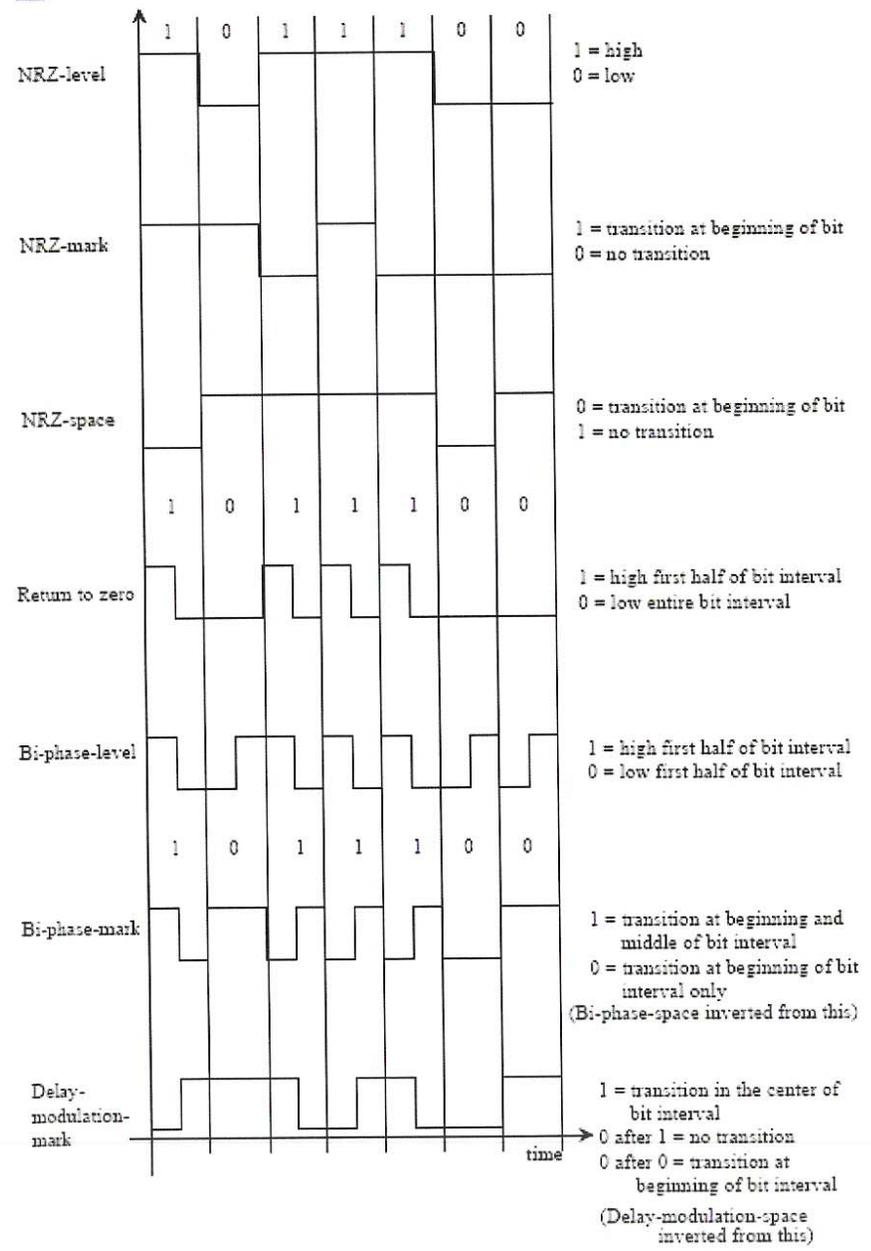
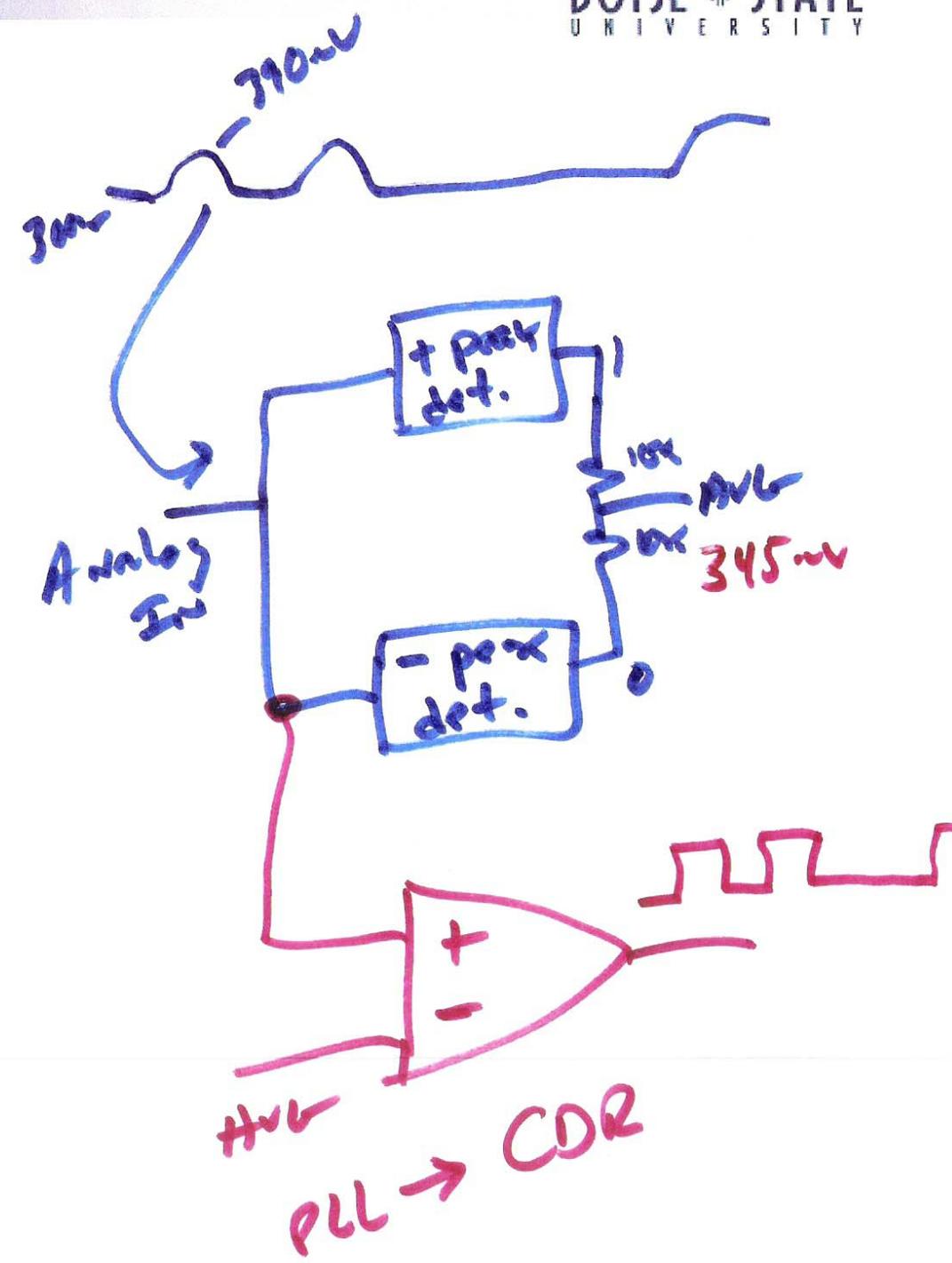


Figure 19.44 Data transmission formats.



6)

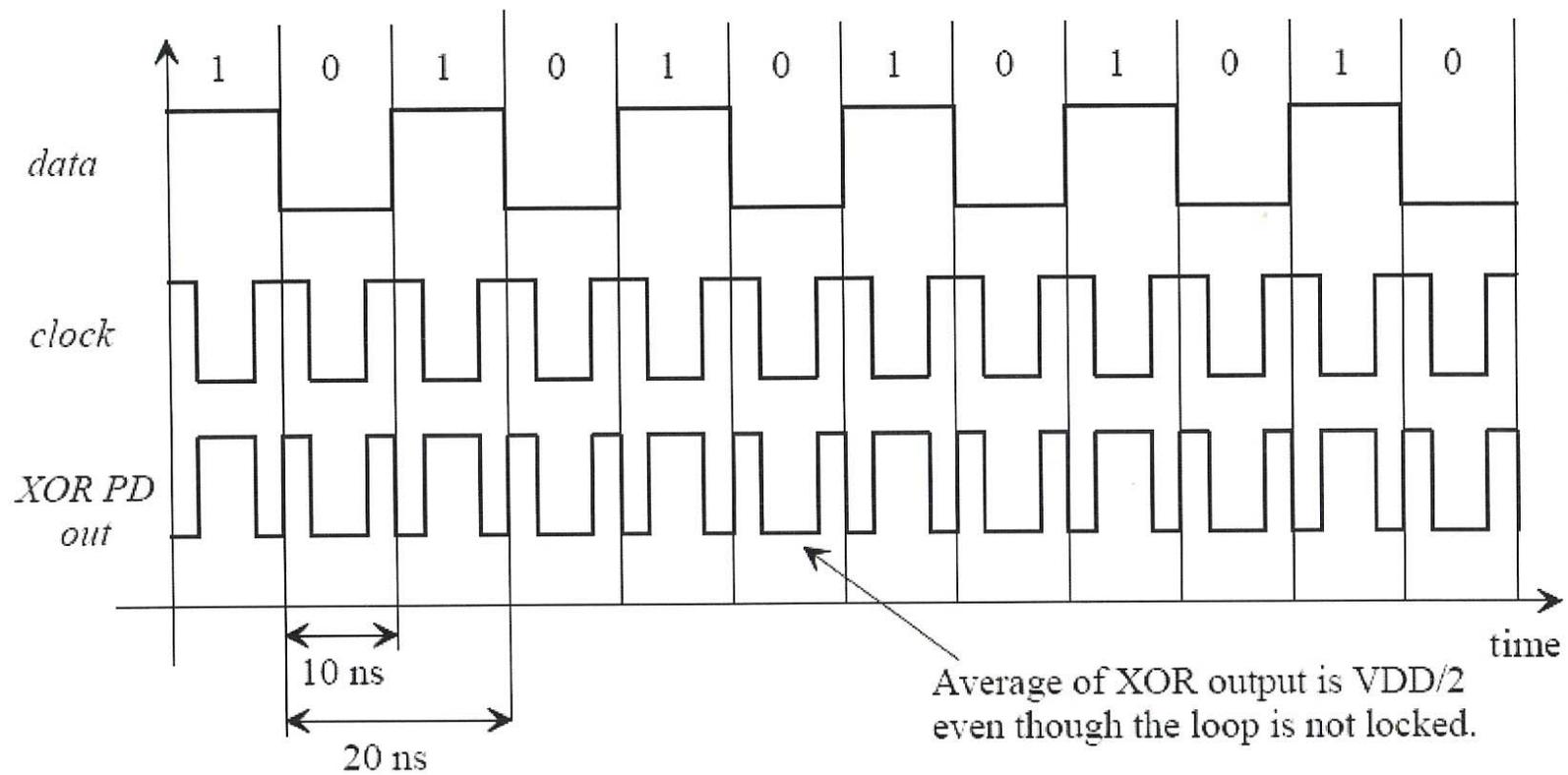


Figure 19.45 The problems of using clock without the divide by 2 to lock on data.

7)

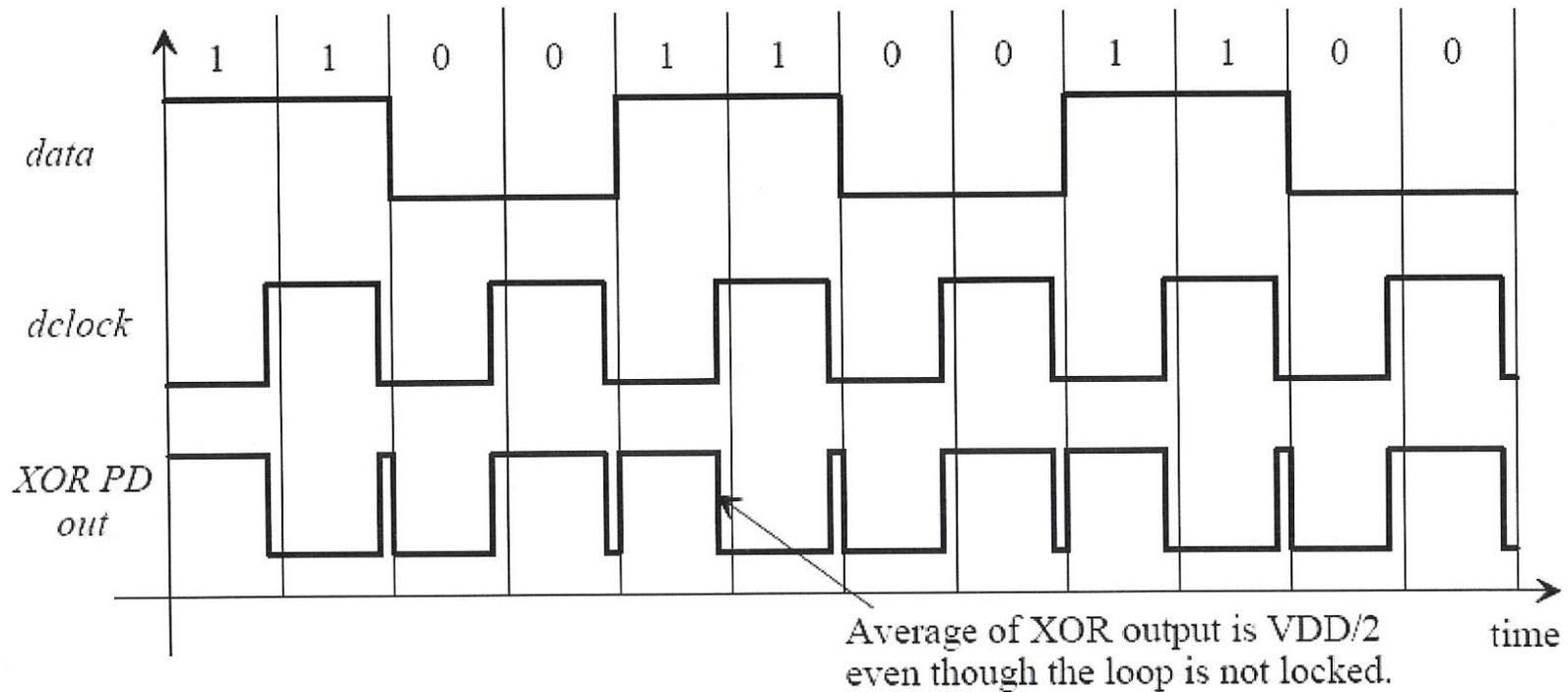


Figure 19.46 Problems trying to lock on a data stream that is one-half the dclock frequency.

8)

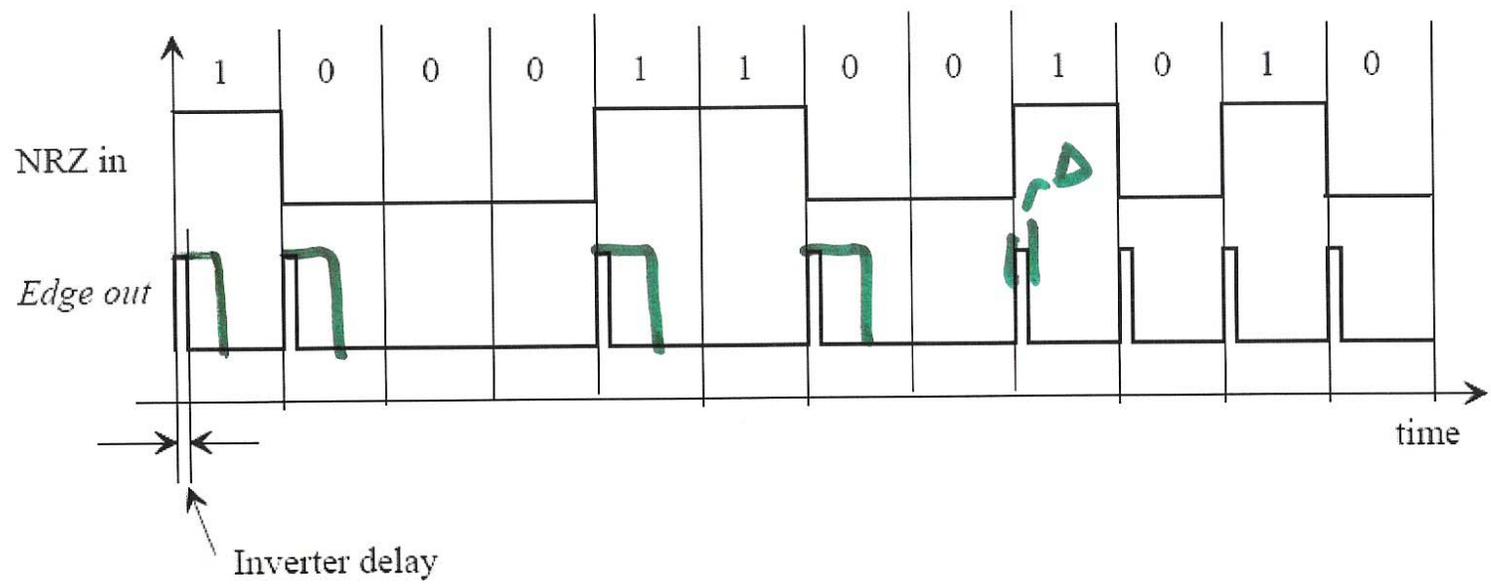
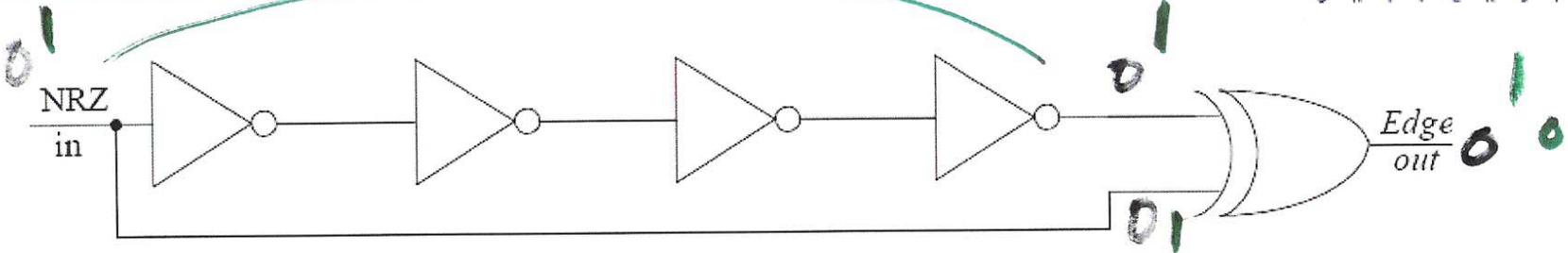
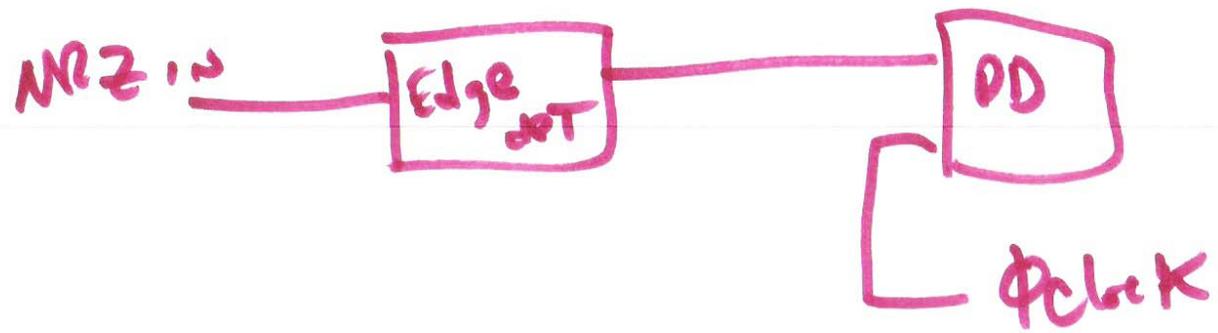
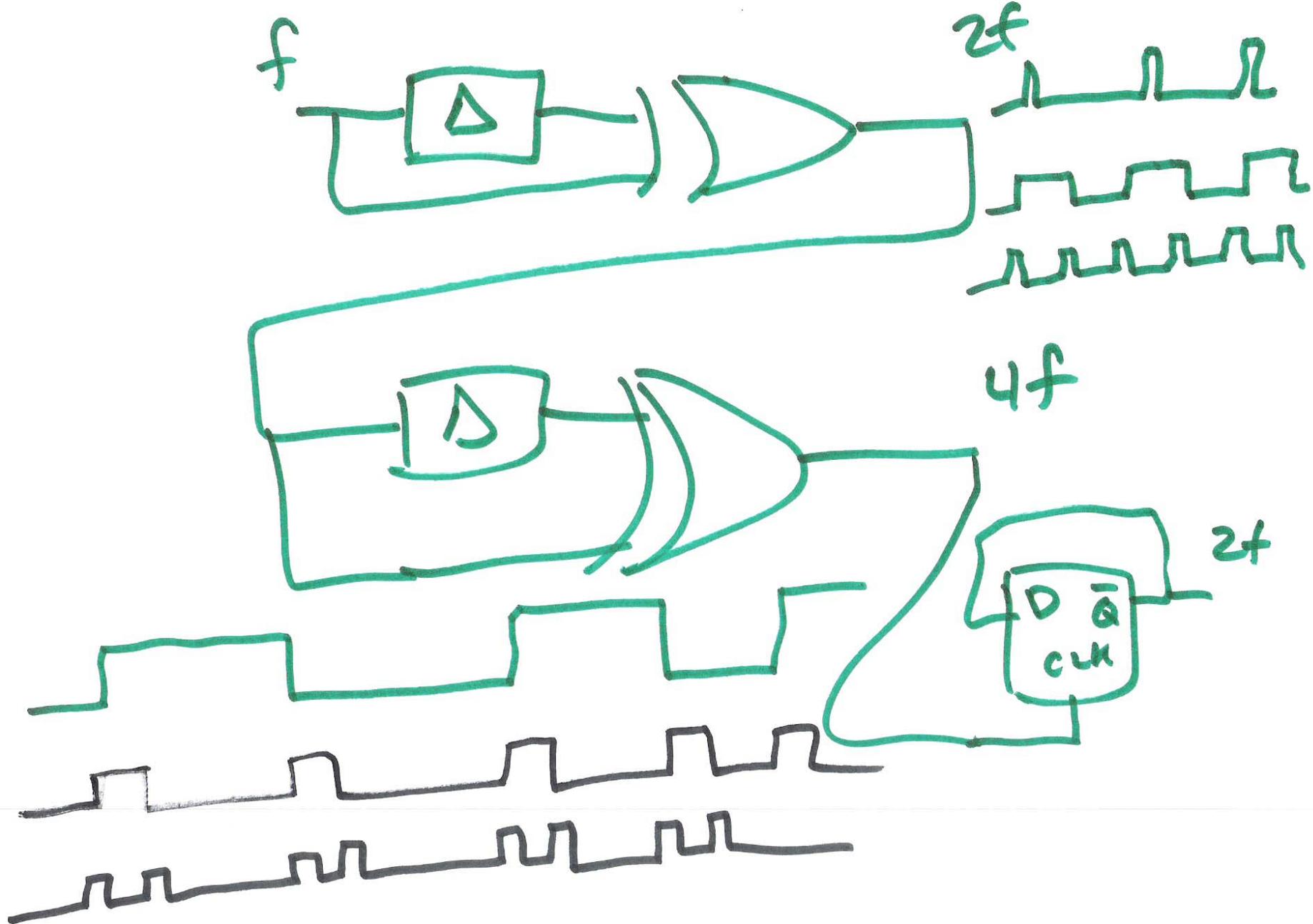


Figure 19.47 Detecting the edges in NRZ data.



9)



10)

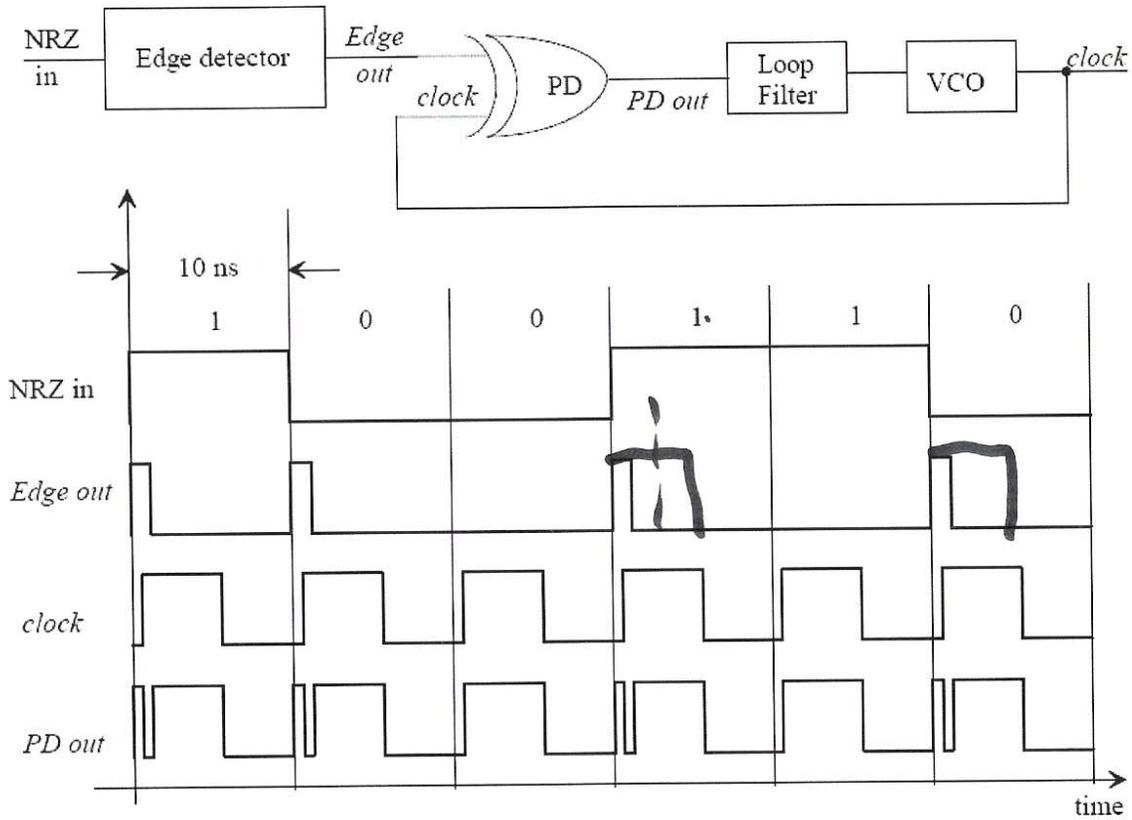
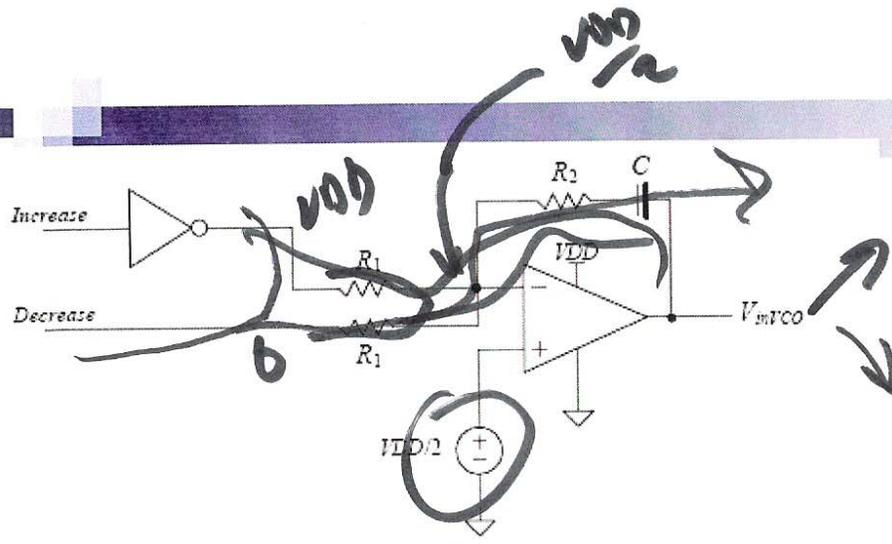
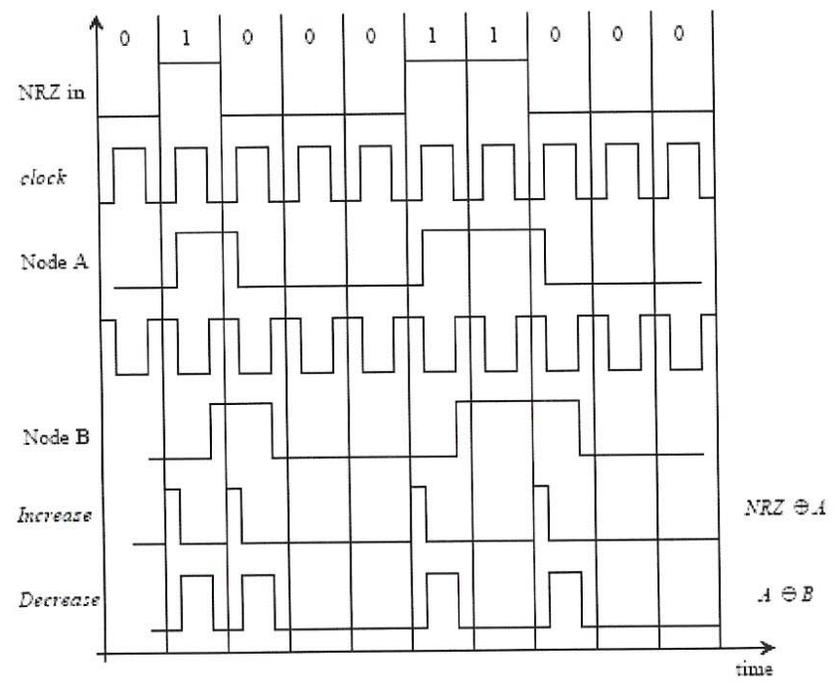


Figure 19.48 Clock-recovery circuit for NRZ using an edge detector. Note that the DPLL is in lock, when the rising edge of clock is centered on the edge output pulse.

11)



(a)



(b)

Figure 19.50 (a) Possible loop filter used in a self-correcting (Hogge) DPLL and (b) waveforms when the loop is not in lock and the clock leads the center of the data.

12)

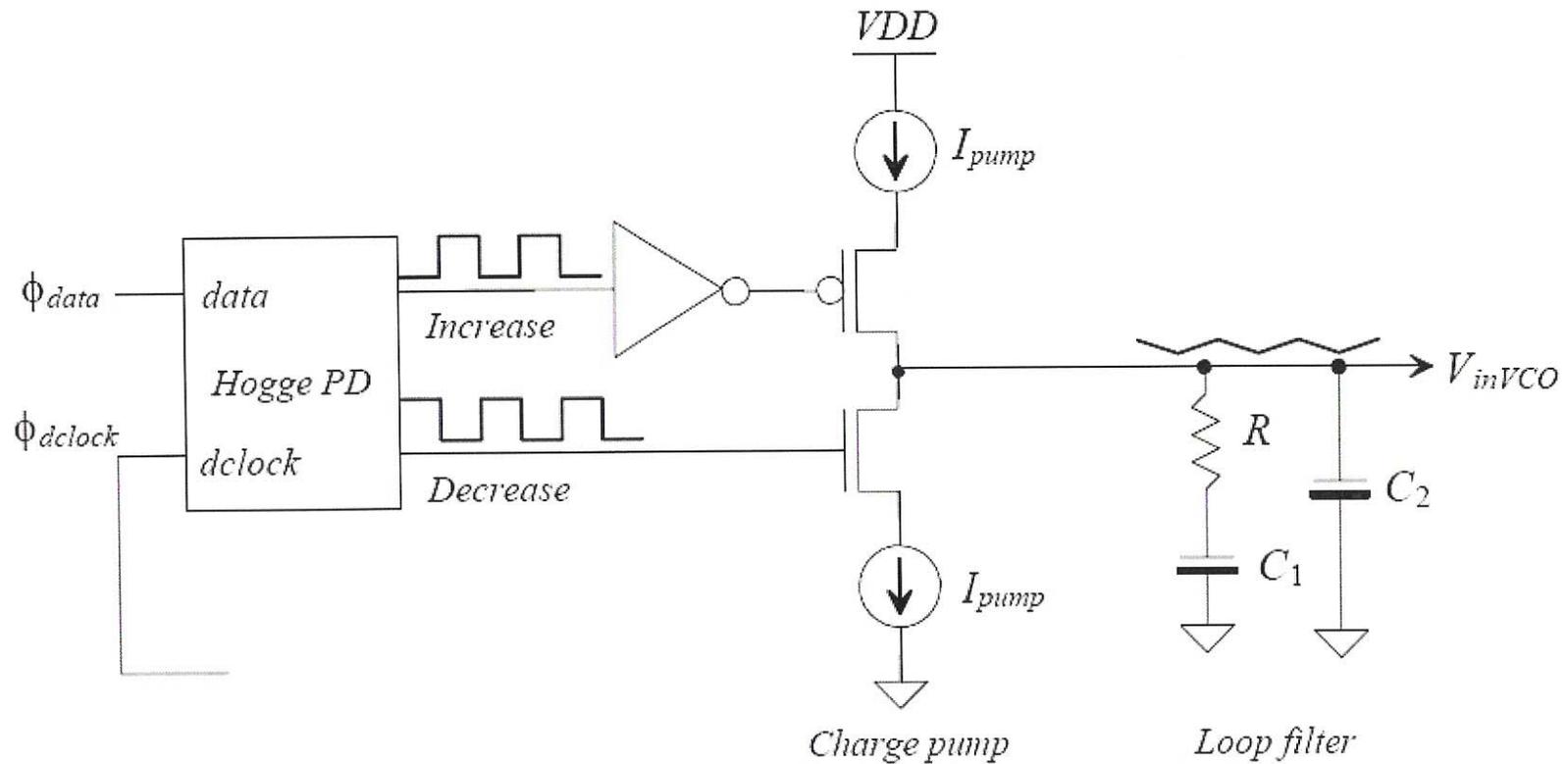


Figure 19.52 Self-correcting PD with charge pump output.

Hogge phase detector

13)